

Summary

- The DTFT of the auto-correlation and of the cross-correlation (extended review, week 02)
 - the DTFT of the auto-correlation
 - the DTFT of the cross-correlation
 - *auto/cross-correlation basic properties*
 - auto/cross-correlation between output and input of an LSI system
 - examples
- The Z-Transform of the auto/cross-correlation (extended review, weeks 04/05)
 - the Z-Transform of the auto-correlation
 - the Z-Transform of the cross-correlation
- Computing the auto/cross-correlation of finite-length sequences using the DFT

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- our starting point
- the cross-correlation using the DFT
- the auto-correlation using the DFT



the DTFT of the auto-correlation

the auto-correlation is defined as (in this discussion, we admit energy signals)

$$r_x[\ell] = x[\ell] * x^*[-\ell] = \sum_{k=-\infty}^{+\infty} x[k] x^*[k-\ell]$$

it characterizes the similarity between a sequence and a copy of itself when it is shifted by a lag (ℓ)

considering the DTFT properties

$$\begin{array}{cccc} x[\ell] & \stackrel{\mathcal{F}}{\longleftrightarrow} & X(e^{j\omega}) \\ x^*[\ell] & \stackrel{\mathcal{F}}{\longleftrightarrow} & X^*(e^{-j\omega}) \\ x[-\ell] & \stackrel{\mathcal{F}}{\longleftrightarrow} & X(e^{-j\omega}) \\ x^*[-\ell] & \stackrel{\mathcal{F}}{\longleftrightarrow} & X^*(e^{j\omega}) \end{array}$$

then

$$r_{x}[\ell] = x[\ell] * x^{*}[-\ell] \quad \stackrel{\mathcal{F}}{\longleftrightarrow} \quad R_{x}(e^{j\omega}) = X(e^{j\omega}) \cdot X^{*}(e^{j\omega}) = |X(e^{j\omega})|^{2}$$

where $R_{\chi}(e^{j\omega}) = |X(e^{j\omega})|^2$ is called the spectral density of energy



- the DTFT of the auto-correlation (cont.)
 - the Wiener-Khintchine Theorem: the auto-correlation and the spectral density of energy form a Fourier pair

$$r_x[\ell] \quad \stackrel{\mathcal{F}}{\longleftrightarrow} \quad R_x(e^{j\omega}) = |X(e^{j\omega})|^2$$

thus,

$$r_{x}[\ell] = \frac{1}{2\pi} \int_{-\pi}^{\pi} R(e^{j\omega}) e^{j\omega\ell} d\omega$$

and, in particular, the energy of the signal can be found using

$$E = r_{x}[0] = \sum_{k=-\infty}^{+\infty} |x[k]|^{2} = \frac{1}{2\pi} \int_{-\pi}^{\pi} R(e^{j\omega}) d\omega = \frac{1}{2\pi} \int_{-\pi}^{\pi} |X(e^{j\omega})|^{2} d\omega$$

which reflects the Parseval Theorem

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• the DTFT of the cross-correlation the cross-correlation is defined as (we admit energy signals)

$$r_{xy}[\ell] = x[\ell] * y^*[-\ell] = \sum_{k=-\infty}^{+\infty} x[k] y^*[k-\ell]$$

it characterizes the similarity between a sequence and a copy of another sequence when it is shifted by a lag (ℓ)

considering the DTFT properties

$$\begin{array}{cccc} x[\ell] & \stackrel{\mathcal{F}}{\longleftrightarrow} & X(e^{j\omega}) \\ y[\ell] & \stackrel{\mathcal{F}}{\longleftrightarrow} & Y(e^{j\omega}) \\ y^*[\ell] & \stackrel{\mathcal{F}}{\longleftrightarrow} & Y^*(e^{-j\omega}) \\ y[-\ell] & \stackrel{\mathcal{F}}{\longleftrightarrow} & Y(e^{-j\omega}) \\ y^*[-\ell] & \stackrel{\mathcal{F}}{\longleftrightarrow} & Y^*(e^{j\omega}) \end{array}$$

then

$$r_{xy}[\ell] = x[\ell] * y^*[-\ell] \quad \stackrel{\mathcal{F}}{\longleftrightarrow} \quad R_{xy}(e^{j\omega}) = X(e^{j\omega}) \cdot Y^*(e^{j\omega})$$



- auto/cross-correlation basic properties
 - complex-conjugate symmetry (auto-correlation only!)

$$r_x[\ell] = r_x^*[-\ell]$$
 $r_{xy}[\ell] = r_{yx}^*[-\ell]$

(implies that $R_x(e^{j\omega})$ is real-valued)

- upper bound

$$|r_x[\ell]| \le r_x[0]$$
 $|r_{xy}[\ell]| \le \sqrt{r_x[0] \cdot r_y[0]}$

- normalized auto-correlation and cross-correlation

$$\rho_{x}[\ell] = \frac{r_{x}[\ell]}{r_{x}[0]}, \quad |\rho_{x}[\ell]| \le 1 \qquad \rho_{xy}[\ell] = \frac{r_{xy}[\ell]}{\sqrt{r_{x}[0] \cdot r_{y}[0]}}, \quad |\rho_{xy}[\ell]| \le 1$$

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- auto/cross-correlation between output and input of an LSI system
 - the (quite important !) last two equations are stated without proof



$$y[\ell] = x[\ell] * h[\ell] \quad \stackrel{\mathcal{F}}{\longleftrightarrow} \quad Y(e^{j\omega}) = X(e^{j\omega})H(e^{j\omega})$$
$$r_{yx}[\ell] = r_x[\ell] * h[\ell] \quad \stackrel{\mathcal{F}}{\longleftrightarrow} \quad R_{yx}(e^{j\omega}) = R_x(e^{j\omega})H(e^{j\omega})$$
$$r_y[\ell] = r_x[\ell] * h[\ell] * h^*[-\ell] \quad \stackrel{\mathcal{F}}{\longleftrightarrow} \quad R_y(e^{j\omega}) = R_x(e^{j\omega})|H(e^{j\omega})|^2$$



let us admit two finite-length discrete-time signals, x[n] and y[n]



it can be easily concluded that

$$\begin{aligned} x[\ell] &= 3\delta[\ell] + 2\delta[\ell-1] + \delta[\ell-2] & \stackrel{\mathcal{F}}{\longleftrightarrow} X(e^{j\omega}) = 3 + 2e^{-j\omega} + e^{-j2\omega} \\ y[\ell] &= \delta[\ell] + 2\delta[\ell-1] + 3\delta[\ell-2] & \stackrel{\mathcal{F}}{\longleftrightarrow} Y(e^{j\omega}) = 1 + 2e^{-j\omega} + 3e^{-j2\omega} \end{aligned}$$

$$R_x(e^{j\omega}) = 3e^{j2\omega} + 8e^{j\omega} + 14 + 8e^{-j\omega} + 3e^{-j2\omega} = R_y(e^{j\omega}), \text{ (why ?)}$$

$$R_{xy}(e^{j\omega}) = 9e^{j2\omega} + 12e^{j\omega} + 10 + 4e^{-j\omega} + e^{-j2\omega}$$



random sequences may exhibit $r_{\chi}[\ell] = \delta[\ell]$



Fig. 1.84 From left to right, the top row of plots represents a few samples of a random sequence (f[n]), the flat PDF of its samples amplitudes, and its auto-correlation function $r_f[\ell]$. The bottom row represents the same sequence of plots in the case the random sequence (g[n]) has a Gaussian PDF. In both cases, the auto-correlation functions, $r_f[\ell]$ and $r_g[\ell]$, are identical and correspond to the unit impulse.



deterministic sequences may exhibit $r_{\chi}[\ell] = \delta[\ell]$



Fig. 1.76 Example of a waveform $x[\ell]$ (top left figure), its auto-correlation function $r_x[\ell]$ (bottom left figure), a waveform $y[\ell]$ consisting of noisy version of $x[\ell+3]$ (top right figure), and the cross-correlation function $r_{xy}[\ell]$ (bottom right figure).



the auto-correlation is useful to find the period of periodic signals: it is signalled by the *first* local maximum in the $\rho_x[\ell]$ or $\rho_x[\ell]$ functions $(\ell \neq 0, \text{ why } ?)$



Fig. 1.80 Panel a) represents a segment of a periodic sequence (x[n]) corresponding to a vowel signal (the sampling frequency is 22050 Hz), and panel c) represents its normalized auto-correlation function $(\rho_x[\ell])$. Panel b) represents a segment of an ECG signal that has been captured using 1 kHz as the sampling frequency. Panel c) represents it normalized auto-correlation function (see text for details on the autocorrelation analysis).



• the Z-Transform of the auto-correlation the auto-correlation is defined as (in this discussion, we admit energy signals)

$$r_{x}[\ell] = x[\ell] * x^{*}[-\ell] = \sum_{k=-\infty}^{+\infty} x[k]x^{*}[k-\ell]$$

considering the Z-Transform properties

$$\begin{array}{ll} x[\ell] & \xleftarrow{Z} & X(z), & RoC = R_x \equiv r_E < |z| < r_D \\ x^*[\ell] & \xleftarrow{Z} & X^*(z^*), & RoC = R_x \\ x[-\ell] & \xleftarrow{Z} & X(z^{-1}), & RoC = 1/R_x \equiv 1/r_D < |z| < 1/r_E \\ x^*[-\ell] & \xleftarrow{Z} & X^*(1/z^*), & RoC = 1/R_x \end{array}$$

then

$$r_{x}[\ell] = x[\ell] * x^{*}[-\ell] \quad \stackrel{Z}{\longleftrightarrow} \quad R_{x}(z) = X(z) \cdot X^{*}(1/z^{*}), \ RoC = R_{x} \cap 1/R_{x}$$

where $R_{\chi}(z) = X(z) \cdot X^*(1/z^*)$ is called the energy spectrum



- the Z-Transform of the auto-correlation (cont.)
 - the Wiener-Khintchine Theorem: the auto-correlation and the energy spectrum form a Z-Transform pair

$$r_x[\ell] \quad \stackrel{Z}{\longleftrightarrow} \quad R_x(z) = X(z) \cdot X^*(1/z^*)$$

thus,

$$r_x[\ell] = \frac{1}{2\pi j} \oint_C R_x(z) Z^{\ell-1} dz$$

and, in particular, the energy of the signal can be found using

$$E = r_x[0] = \sum_{k=-\infty}^{+\infty} |x[k]|^2 = \frac{1}{2\pi j} \oint_C X(z) \cdot X^*(1/z^*) Z^{-1} dz$$

which reflects the Parseval Theorem in the Z-domain

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• the Z-Transform of the cross-correlation the cross-correlation is defined as (we admit energy signals)

$$r_{xy}[\ell] = x[\ell] * y^*[-\ell] = \sum_{k=-\infty}^{+\infty} x[k] y^*[k-\ell]$$

considering the Z-Transform properties

then

$$r_{xy}[\ell] = x[\ell] * y^*[-\ell] \xleftarrow{Z} R_{xy}(z) = X(z) \cdot Y^*(1/z^*), \ RoC = R_x \cap 1/R_y$$
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- Computing the AC/CC of finite-length sequences using the DFT
- our starting point
 - we have seen that if a sequence x[n] has length M, and another sequence h[n] has length L, the linear convolution between them corresponds to a sequence whose length is L+M-1
 - we also have seen that if the signals are zero-padded and made periodic with period N, then the linear result convolution result may also be found using the DFT and its properties as long as $N \ge L+M-1$
 - in this case, the circular convolution yields the same result of the linear convolution according to the following block diagram





- the cross-correlation using the DFT
 - assuming that both sequences, $x[\ell]$ and $y[\ell]$, are suitably zeropadded such that the circular convolution reduces to the linear convolution, then $r_{xy}[\ell]$ can be computed using DFT-based frequency domain processing





- the auto-correlation using the DFT
 - this case is a particular case in the sense that $x[\ell] = y[\ell]$ in the previous slide, which leads to the following (simplified) relationships and block diagram that yields $r_x[\ell]$

