

## L.EEC/M.EEC L.EEC025 - Fundamentals of Signal Processing

## FIRST EXAM, FEBRUARY 01, 2024 Duration: 120 Minutes, closed book

**NOTE**: each question *must* be answered in a separate sheet; please write your name and Student order number on all sheets, please provide complete answers while trying to minimize paper usage.

1. Three causal discrete-time systems have the illustrated zero-pole diagrams A, B, and C, and the illustrated frequency response magnitudes 1, 2, and 3. Admit that the radius of all poles and zeros is either 0.8 or 1/0.8, and that the angles of all poles and zeros are multiples of  $\pi/4$  rad.



- a) [1,5 *pts*] Match each zero-pole diagram (A, B, C) to the corresponding frequency response magnitude (1, 2, 3), and indicate the main supporting arguments.
- b) [1 *pt*] Which of the represented systems (A, B, or C) are stable ? And which ones are minimum-phase systems ? Why ?
- c) [1 *pt*] Consider the new system formed by cascading systems A and B. What is the order of the new system ? Is the new system IIR or FIR ? Is it a linear-phase system ?
- d)  $[1 pt] \ll If H_A(z), H_B(z), and H_C(z)$  represent the transfer function of systems A, B and C, respectively, then  $H_A(z)$  may be obtained by combining  $H_B(z)$  and  $H_C(z)$  in a suitable way». Is this statement true or false? If true, what is the suitable combination?
- 2. Consider system C as described in Prob. 1.
  - a) [1,5 pts] Find the transfer function of the system, H(z), write a difference equation implementing it and sketch a corresponding canonic realization structure.

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- b) [1,5 *pt*] Obtain a compact expression characterizing the magnitude of the frequency response of the system,  $|H(e^{j\omega})|$ , and show that its maximum gain depends on  $\sqrt{1 + 2 \times 0.8^{-4} + 0.8^{-8}}$ , and its minimum gain depends on  $\sqrt{1 2 \times 0.8^{-4} + 0.8^{-8}}$ .
- c) [1 *pt*] Consider the illustrated analog and causal discrete-time system whose frequency response is as suggested in **b**). The sampling frequency is 100 Hz. The analog input signal is  $x_c(t) = \sin(175\pi t) + \sin(450\pi t)$ . Notice that an *anti-aliasing* filter does not exist.



Find the sinusoidal frequencies of the discrete-time signal x[n] in the Nyquist range, i.e. in the range  $-\pi \le \omega < \pi$ . Obtain a compact expression for x[n].

- d) [1 *pt*] Presuming ideal reconstruction conditions, indicate what sinusoidal frequencies (in Hertz) exist in  $y_c(t)$ , and indicate what their magnitudes are.
- 3. Consider the following Matlab code.

```
x=[4j 0 0 0 1 2j 3];
N=length(x);
X=fft(x); Y=zeros(1,N);
Y(1)=X(1); Y(N:-1:2)=X(2:N);
Z=conj(Y).*Y;
z=ifft(Z);
```

- a) [1 *pt*] Without executing the code, find the result of ifft(Y).
- **b)** [1,5 *pts*] Without executing the code, find the result of ifft(Z).
- c) [1 *pt*] Without executing the code, explain why z(1) = sum(x.\*conj(x)) and explain why the z[n] sequence is conjugate-symmetric.
- 4. Figure A represents the fundamental frequency (or pitch, in Hz) obtained from the singing of a female singer. The signal in Fig. A is available in vector vbt whose samples presume a sampling period of  $\frac{512}{22050} \approx 23.22$  ms. The oscillation of the pitch around its average value (about 388.5 Hz) is called vibrato. In this exercise, we want to find the vibrato frequency using DFT analysis. Figures B and C represent the absolute value of the gain-normalized DFT using the following Matlab code in one case:

X=fft(vbt, N); % N=46, N2= N/2
plot([0:N2-1], abs(X(1:N2))/sum(vbt))

and using the alternative Matlab code in the other case:

X = fft(vbt-mean(vbt)); % N=46, N2= N/2
plot([0:N2-1], abs(X(1:N2))/sum(vbt))

Both data tips in Figure B and C denote a local maximum for k = 6. Using an accurate frequency estimator, the local maximum on the DFT bin scale is found for  $k \approx 5.6$ .



- a) [1 *pt*] Explain which code generates which figure, and why.
- b) [1,5 *pts*] Estimate the vibrato frequency in Hertz. Explain your reasoning.
   NOTE: the typical vibrato frequency is between 5 Hz and 8 Hz.
- c) [1 pt] We admit that the signal represented in Fig. A is real-valued. If prior to FFT transformation the signal is multiplied by a Hamming window, instead of the (default) Rectangular window, what is the expected impact of that on |X[k]|, and on the accuracy of the vibrato frequency estimation ?
- 5. Admit that you use an FIR adaptive filter of length 16 (i.e., the order is 15) in a system identification configuration. The 'black box' system consists of a second order all-pass system whose zero-pole diagram is represented in Figure A, and the adaptive filter starts from rest (i.e., all coefficients are zero at start). The sampling frequency is 8 kHz, the excitation is white Gaussian noise, adaptation is implemented using the LMS algorithm  $(h_{n+1}[k] = h_n[k] + \mu e[n]x[n-k])$  and various adaptation step sizes ( $\mu$ ).

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- a) [1,25 *pts*] Sketch a block diagram of the adaptive system that is configured as system identification. Describe its main blocks.
- b) [1,25 *pts*] The Root Mean Square Error (RMSE) of the adaption process is represented on the right-hand side for various values of  $\mu$ : -0.025, 0.025, 0.05, 0.5. Based on the RMSE trends, what value do  $\mu_A$ ,  $\mu_B$ ,  $\mu_C$  and  $\mu_D$  represent ? Why?
- c) [1 *pt*] After adaptation, the identified system has the zero-pole diagram represented in Figure B. How do you explain that the system in Figure A is identified as the system in Figure B?



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